

7 Inter-symbol Interference and Pulse Shaping

7.1. Recall that, in *Pulse-Amplitude Modulation* (PAM), we start with a sequence of numbers

$$\dots, m[-3], m[-2], m[-1], m[0], m[1], m[2], m[3], \dots$$

discrete-time message

as shown in Figure 36.

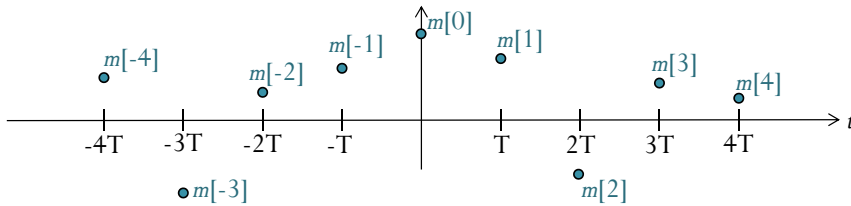


Figure 36: Sequence of Numbers for PAM

- These numbers may come from sampling a continuous-time signal $m(t)$. Alternatively, it may directly represent (digital) information that intrinsically available in discrete-time.
- Because the $m[n]$ may not come from sampling, we call each $m[n]$ a **symbol**.

We use these numbers to modify (modulate) the height (amplitude) of a pulse train. A single pulse is denoted by $p(t)$. This pulse occurs every T seconds.

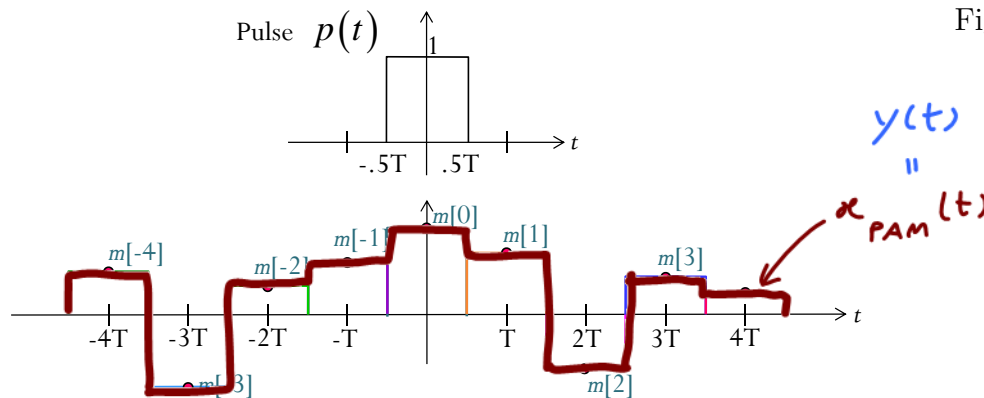
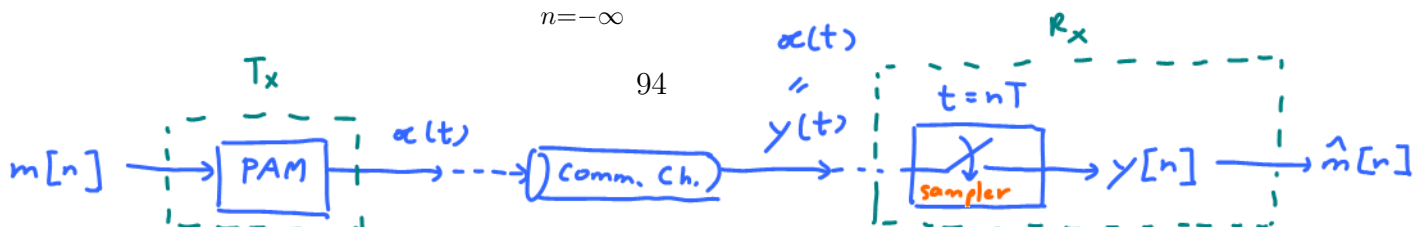


Figure 37: PAM

The PAM signal is then

$$x(t) = \sum_{n=-\infty}^{\infty} m[n] p(t - nT)$$



This signal is transmitted via the communication channel which usually corrupts it. At the receiver, the received signal is $y(t)$. A more advanced receiver would try to first cancel the effect of the channel. However, for simplicity, let's assume that our receiver simply samples $y(t)$ every T seconds to get

$$y[n] = y(t)|_{t=nT}$$

and we will take this to be the estimate $\tilde{m}[n]$ of our $m[n]$.

- If $m[n]$ is the sampled version of $m(t)$, then at the receiver, after we recover $m[n]$, we can reconstruct $m(t)$ by using the reconstructing equation (58).

Because our assumed receiver is so simple, we are going to also assume²⁵ that $y(t) = x(t)$.

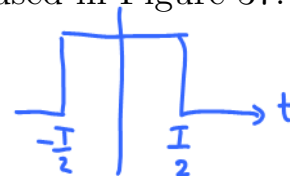
7.2. In this section, our goal is to design a “good” pulse $p(t)$ that satisfies two important properties

- $\tilde{m}[n] = m[n]$ for all n . Under our assumptions above, this means we want $x[n] \equiv x(nT) = m[n]$ for all n .
- $P(f)$ is band-limited and hence $X(f)$ is band-limited.

We will first give examples of “poor” $p(t)$.

Example 7.3. Let's consider the rectangular pulse used in Figure 37:

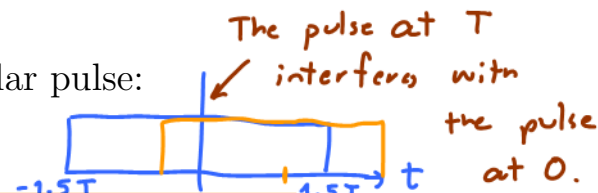
$$p(t) = 1[|t| \leq T/2].$$



- $\tilde{m}[n] = m[n]$ for all n .
- The Fourier transform of the rectangular pulse is a sinc function. So, it is not band-limited.

Example 7.4 (Slide). Let's try a wider rectangular pulse:

$$p(t) = 1[|t| \leq 1.5T].$$



Here, we face a problem called **inter-symbol interference (ISI)** in our sequence $\tilde{m}[n]$ at the receiver. The pulses are too wide; they interfere with other pulses at the sampling time instants (decision making instants), making $\hat{m}[n] \neq m[n]$.

²⁵Alternatively, we may assume that there is an earlier part of the receiver that (perfectly) eliminates the effect of the channel for us.

Example 7.5 (slide). $p(t) = 1[|t| \leq T/4]$.

- When the pulse $p(t)$ is narrower than T , we know that the pulses in PAM signal will not overlap and therefore we won't have any ISI problem.

Example 7.6 (slides).

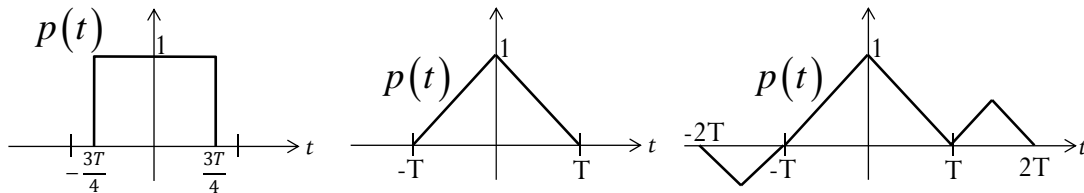


Figure 38: Examples of pulses that do not cause ISI.

- Even when the pulses are wider than T , if they do not interfere with other pulses at the sampling time instants (decision making instants), we can still have no ISI.

7.7. We can now conclude that a “good” pulse satisfying condition (a) in 7.2 must **not cause inter-symbol interference (ISI)**: at the receiver, the n th symbol $\tilde{m}[n]$ should not be affected by the preceding or succeeding transmitted symbol $m[k]$, $k \neq n$. This requirement means that a “good” pulse should have the following property:

$$p[n] \equiv p(t)|_{t=nT} = \begin{cases} 1, & n = 0, \\ 0, & n \neq 0. \end{cases}$$

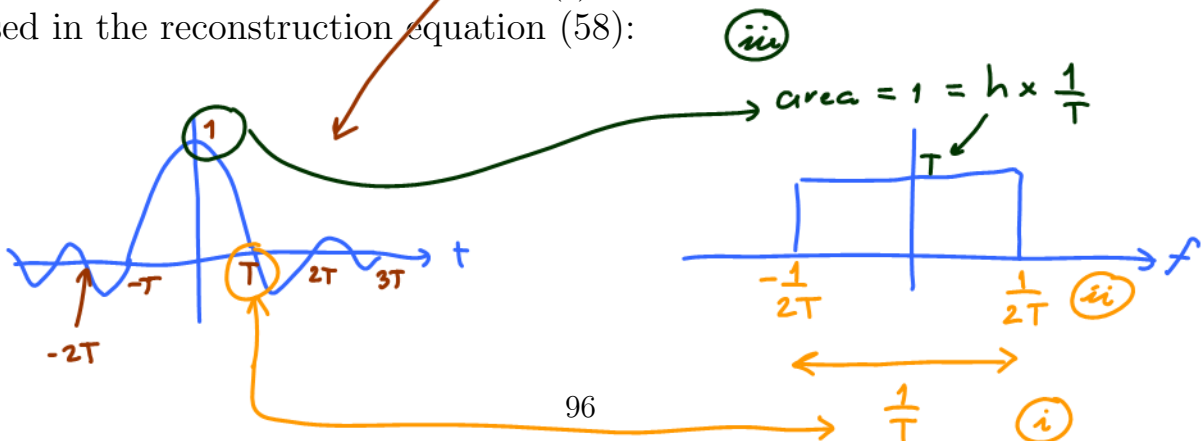
(criterion) first form of the requirement

↑ in the time domain

$\delta[n]$ ← discrete-time version of the Dirac delta func. (59)

Combining this with condition (b) in 7.2, we then want “band-limited pulses specially shaped to avoid ISI (by satisfying (59))” [3, p 506].

7.8. An obvious choice for such $p(t)$ would be the sinc function that we used in the reconstruction equation (58):



Recall Figure 33, repeated here (with modified labels) as Figure 39.

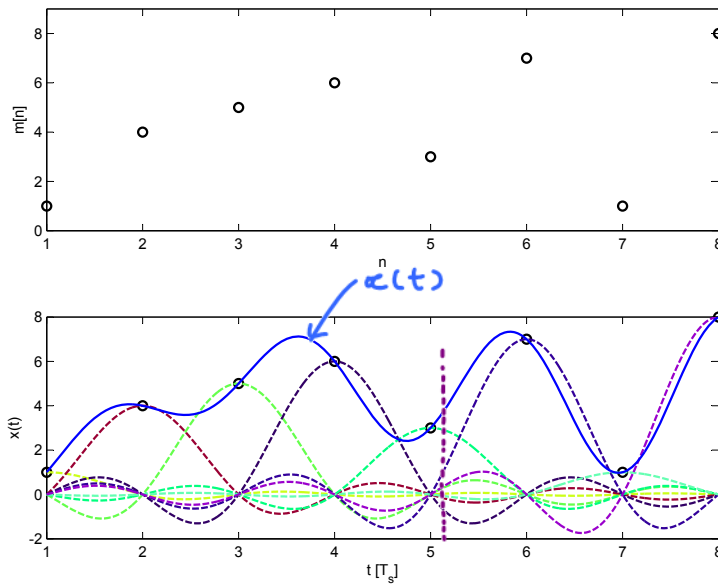


Figure 39: Using the sinc pulse in PAM

Practically, there are problems that force us to seek better pulse shape.

- (a) Infinite duration (Does not decay fast enough)
- (b) Steep slope at each 0-intercept. \Rightarrow sensitive to timing/sync. error
- (c) $\max_t \{x(t)\}$ could be a lot larger than $\max_n \{m[n]\}$.

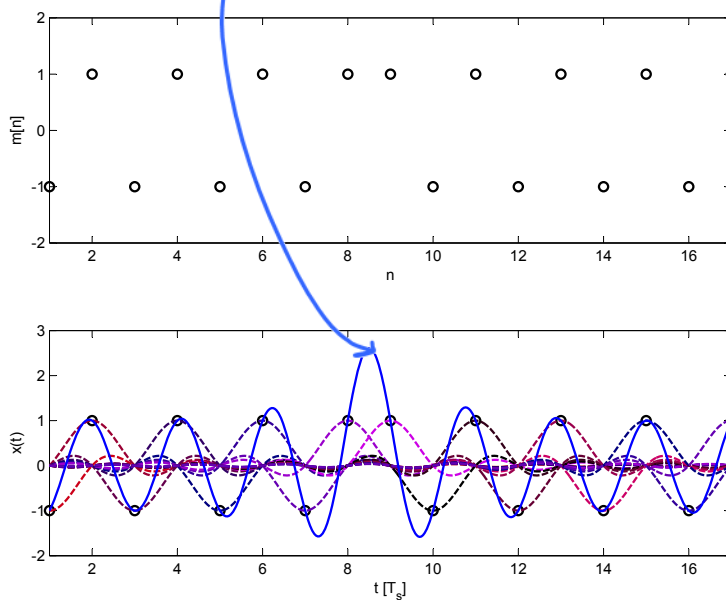


Figure 40: Using the sinc pulse in PAM can cause high peak.

7.9. Because the sinc function may not be a good choice, we now have to consider other pulses that are band-limited and also satisfy (59). To check that a signal is band-limited, we need to look in the frequency domain. However, condition (59) is specified in the time domain. Therefore, we will try to translate condition (59) into a requirement in the frequency domain.

7.10. Note that condition (59) considers $p(t)|_{t=nT}$ which can be thought of as the samples $p[n]$ of the pulse $p(t)$ where the sampling period is $T_s = T$. Recall, from (57), that

$$g_\delta(t) = \sum_{n=-\infty}^{\infty} g[n] \delta(t - nT_s) \xleftrightarrow[\mathcal{F}^{-1}]{\mathcal{F}} G_\delta(f) = f_s \sum_{k=-\infty}^{\infty} G(f - kf_s). \quad f_s = \frac{1}{T_s}$$

Therefore,

$$p_\delta(t) = \sum_{n=-\infty}^{\infty} p[n] \delta(t - nT) \xleftrightarrow[\mathcal{F}^{-1}]{\mathcal{F}} P_\delta(f) = \frac{1}{T} \sum_{k=-\infty}^{\infty} P\left(f - \frac{k}{T}\right). \quad (60)$$

$\delta[n] = \begin{cases} 1, & n=0, \\ 0, & n \neq 0. \end{cases}$

On the LHS, by condition (59), the only nonzero term in the sum is the one with $n = 0$. Therefore, condition (59) is equivalent to $p_\delta(t) = \delta(t)$. However, recall that $\delta(t) \xleftrightarrow[\mathcal{F}^{-1}]{\mathcal{F}} 1$. Therefore, we must have $P_\delta(f) \equiv 1$. Hence, to check condition (59), we can equivalently check that the RHS of (60) must be $\equiv 1$.

Note that $P_\delta(f)$ is “periodic” (in the freq. domain) with “period” $\frac{1}{T}$. (Recall that $G_\delta(f)$ is “periodic” (in the freq. domain) with “period” f_s .) Therefore, the checking does not need to be performed across all frequency f . We only need to focus on one period: $|f| \leq \frac{1}{2T}$.

This observation is formally stated as the “Nyquist’s criterion” below.

7.11. Nyquist’s (first) Criterion for Zero ISI: A pulse $p(t)$ whose Fourier transform $P(f)$ satisfies the criterion

new condition ① $\rightarrow \sum_{k=-\infty}^{\infty} P\left(f - \frac{k}{T}\right) \equiv T, \quad |f| \leq \frac{1}{2T}$ (61)

has sample values satisfying condition (59):

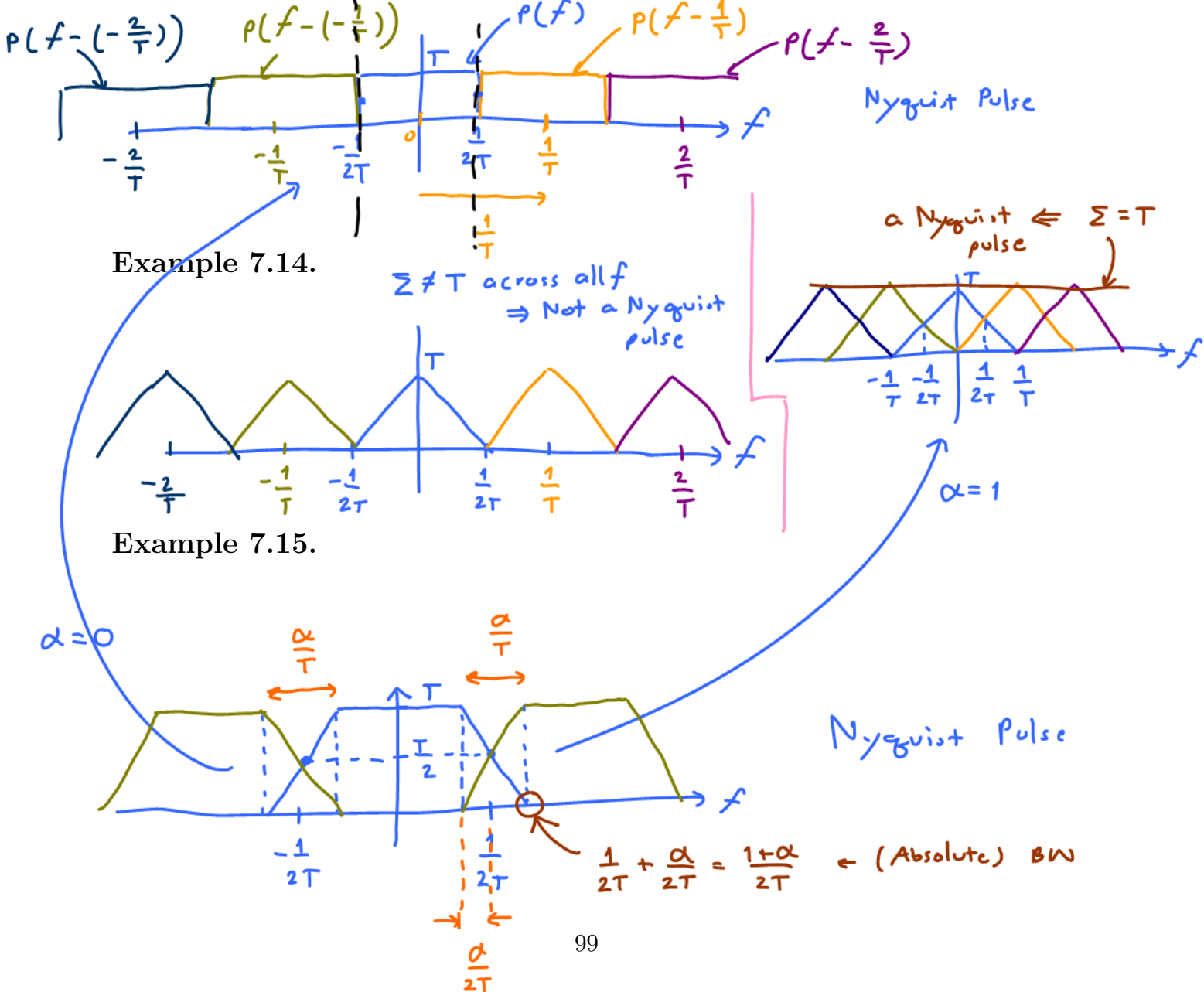
$$p[n] = p(t)|_{t=nT} = \begin{cases} 1, & n = 0, \\ 0, & n \neq 0. \end{cases}$$

- Using this pulse, there will be no ISI in the sample values of

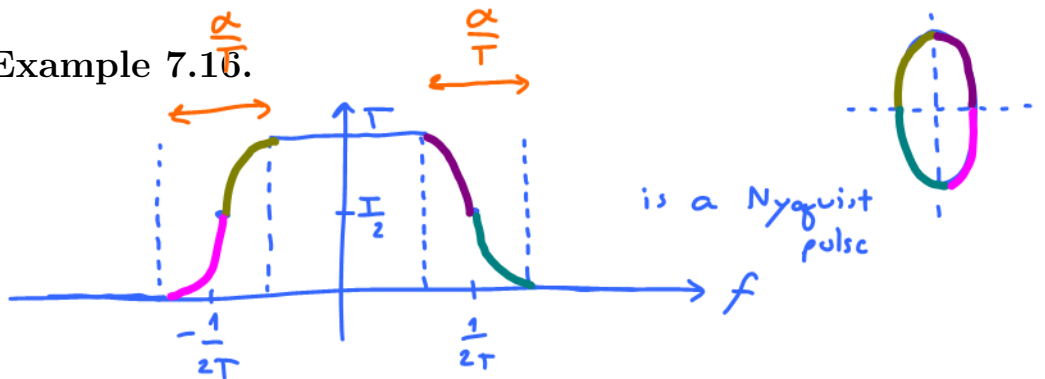
$$y(t) = \sum_{n=-\infty}^{\infty} m[n] p(t - nT)$$

Definition 7.12. A pulse $p(t)$ is a **Nyquist pulse** if its Fourier transform $P(f)$ satisfies (61) above.

Example 7.13. We know that the sinc pulse we used in Example 7.8 works (causing no ISI). Let's check it with the Nyquist's criterion:



Example 7.16.



Example 7.17. An important family of Nyquist pulses is called the **raised cosine** family. Its Fourier transform is given by

$$P_{RC}(f; \alpha) = \begin{cases} T, & 0 \leq |f| \leq \frac{1-\alpha}{2T} \\ \frac{T}{2} \left(1 + \cos \left(\frac{\pi T}{\alpha} \left(|f| - \frac{1-\alpha}{2T} \right) \right) \right), & \frac{1-\alpha}{2T} \leq |f| \leq \frac{1+\alpha}{2T} \\ 0, & |f| \geq \frac{1+\alpha}{2T} \end{cases}$$

with a parameter α called the **roll-off factor**.

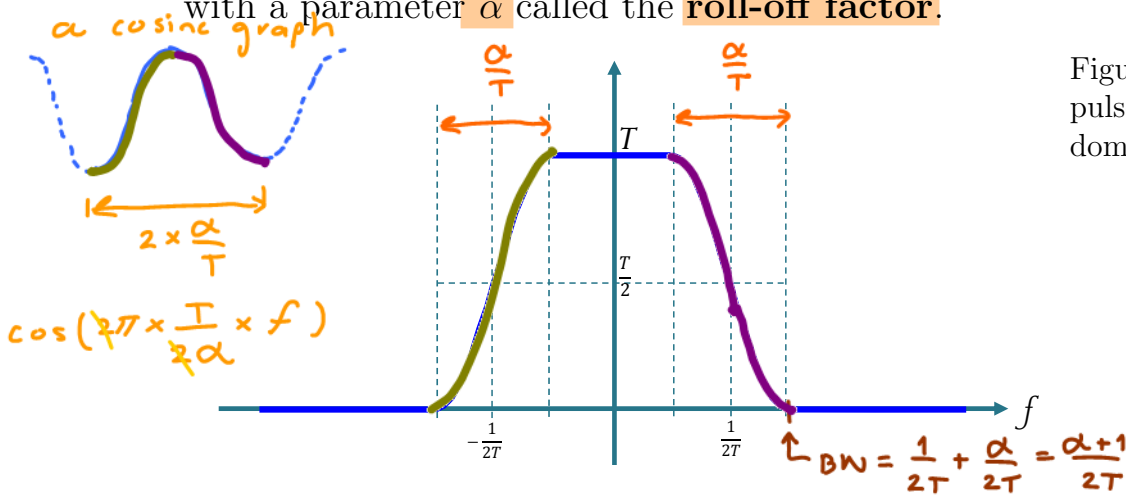


Figure 41: Raised cosine pulse (in the frequency domain)

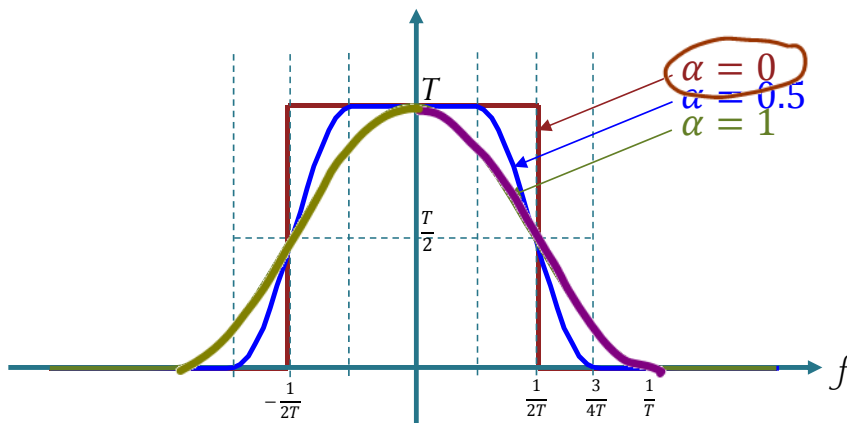


Figure 42: Raised cosine pulse (in the frequency domain) with different values of the roll-off factor

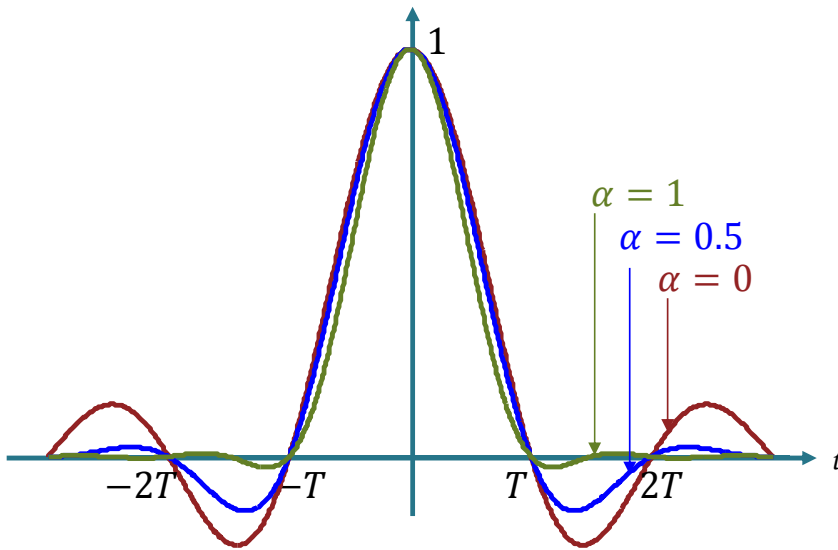
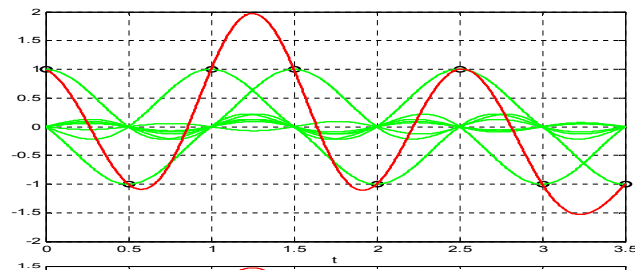


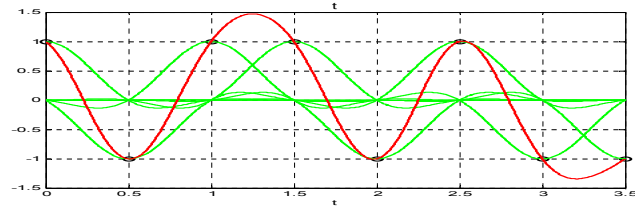
Figure 43: Raised cosine pulse (in the time domain) with different values of the roll-off factor

$$x(t) = \sum_{n=-\infty}^{\infty} m[n] p(t-nT)$$

a) $p(t) = p_{RC}(t;0)$



b) $p(t) = p_{RC}(t;0.5)$



c) $p(t) = p_{RC}(t;1)$

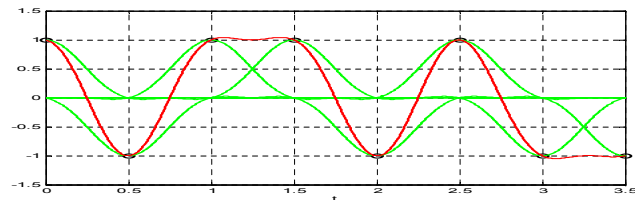


Figure 44: Using the raised cosine pulses in PAM